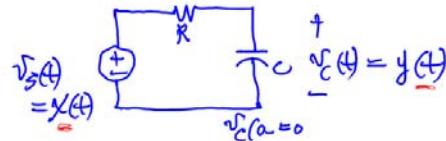


EE103 Lect #7 Oct 13, 2017

Note: Course materials (lecture slides, HW & solutions) are posted on EE103 website:
<https://ee103-fall2017-01.courses.soe.ucsc.edu>

Also webcast can be watched at
<https://webcast.ucsc.edu/EE103>

Let us consider a simple RC circuit

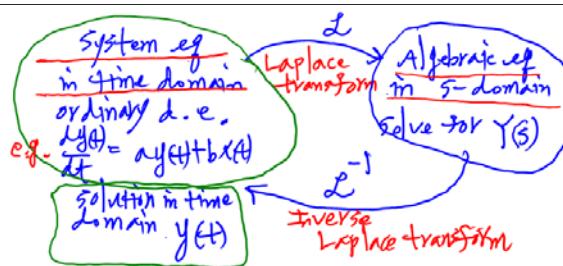


$$\dot{v}_c(t) = C \frac{dy(t)}{dt} = C \frac{dy}{dt} \quad (1)$$

$$\dot{v}_c(t) = \dot{v}_R(t) = \frac{v_s(t) - v_c(t)}{R} = \frac{x(t) - y(t)}{R} \quad (2)$$

From (1) & (2)

$$C \frac{dy}{dt} = \frac{1}{R} (x - y) \text{ or } \boxed{RC \frac{dy}{dt} + y = x} \quad (3)$$



Taking Inverse Laplace Transform of (6) \rightarrow

$$y(t) = \mathcal{L}^{-1}\left(\frac{1}{RCs+1}\right) = \frac{1}{RC} e^{-\frac{1}{RC}t} u(t)$$

$$y(t) = \frac{1}{RC} e^{-\frac{1}{RC}t}$$

$$y(t) = \begin{cases} 0 & t < 0 \\ \frac{1}{RC} e^{-\frac{1}{RC}t} & t \geq 0 \end{cases}$$

$$\begin{aligned} v_s(t) &= \delta(t) \\ x(t) &= \mathcal{L}^{-1}(Y(s)) = \delta(t) \end{aligned}$$

$$y(t) = \frac{1}{RC} \delta(t) e^{-\frac{1}{RC}t}$$

Input Response

$$x(t) = \delta(t) \quad \boxed{x = \frac{d}{dt} y} \quad y(t) = \frac{1}{RC} u(t) e^{-\frac{1}{RC}t} \quad \boxed{\text{impulse response}}$$

For arbitrary $x(t)$ what would be $y(t) = ?$



$$(\text{Ans}) \quad y(t) = x(t) * h(t)$$

WHY?

$$x(t) = \sum_{n=-\infty}^{\infty} x(n) \text{rect}(t-nT) \quad x(t) = \sum_{n=-\infty}^{\infty} [u(t-nT) - u(t-(n+1)T)]$$

$$\begin{aligned} x(t) &= x(0) \text{rect}(t|T) + x(1) \text{rect}(t-T|T) \\ &\quad + x(2T) \text{rect}(t-2T|T) \\ &\quad + x(3T) \text{rect}(t-3T|T) \\ &\quad + x(4T) \text{rect}(t-4T|T) \end{aligned}$$

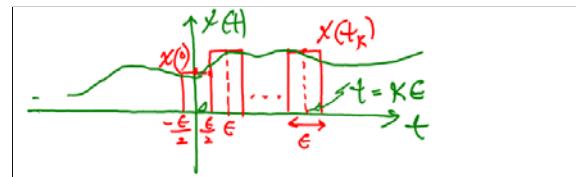
$$= \sum_{k=0}^N x(k\epsilon) \frac{1}{\epsilon} \text{rect}(t-k\epsilon | \epsilon) e$$

$$\hat{x}(t) = \lim_{\epsilon \rightarrow 0} \sum_{k=0}^{\infty} x(k\epsilon) \frac{1}{\epsilon} \text{rect}(t-k\epsilon | \epsilon) e$$

$\overline{\epsilon}$
 $k\epsilon = \tau$
 $\epsilon = dt$
 $N \rightarrow \infty$

$$= \int_0^\infty x(\tau) s(t-\tau) d\tau$$

$$= \hat{x}(t) \cdot 1 = x(t)$$



We note that $x(t)$ can be expressed as

$$x(t) = \lim_{\epsilon \rightarrow 0} \sum_{k=-\infty}^{\infty} x(k\epsilon) \frac{1}{\epsilon} \text{rect}(t-k\epsilon | \epsilon) e$$

$$= \int_{-\infty}^{\infty} x(\tau) s(t-\tau) d\tau$$

Also $y(t) = H[x(t)]$

$$= H \left[\lim_{\epsilon \rightarrow 0} \sum_{k=0}^{\infty} x(k\epsilon) \frac{1}{\epsilon} \text{rect}(t-k\epsilon | \epsilon) e \right]$$

$$= \lim_{\epsilon \rightarrow 0} \sum_{k=0}^{\infty} x(k\epsilon) H \frac{1}{\epsilon} \text{rect}(t-k\epsilon | \epsilon) e$$

$$= \int_0^{\infty} x(\tau) h(t-\tau) d\tau$$

$$= x(t) * h(t) = h(t) * x(t)$$

convolution

Back to the simple RC circuit

What is $y(t) \rightarrow x(t) = u(t) e^{-\frac{t}{RC}}$

$$y(t) = x(t) * h(t) = \int_0^{\infty} u(\tau) e^{-\frac{t-\tau}{RC}} \cdot u(t-\tau) d\tau$$

$$= \int_0^t u(\tau) e^{-\frac{t-\tau}{RC}} u(t-\tau) d\tau$$

$$= \int_0^t \frac{1}{RC} e^{-\frac{t-\tau}{RC}} d\tau$$

$$= \int_0^t d[-e^{-\frac{t-\tau}{RC}}] = y(t)$$

$$= (1 - e^{-\frac{t}{RC}}) = y(t)$$

$$y(t) = 1 - e^{-\frac{t}{RC}}$$

Circuit diagram of a simple RC series circuit with voltage source u(t), resistor R, and capacitor C. The output voltage $y(t)$ is across the capacitor.

$$v_{SC} = x(t) = u(t)$$

$$v_C(0) = 0$$

$$v_C(t) = y(t) = 1 - e^{-\frac{t}{RC}}$$

Convolution of periodic signal

Let $h(t)$ be the triangular pulse shown in Fig. 2-10(a) and let $x(t)$ be the unit impulse train [Fig. 2-10(b)] expressed as

$$x(t) = \delta_T(t) = \sum_{n=-\infty}^{\infty} \delta(t - nT) \quad (2.68)$$

Determine and sketch $y(t) = h(t) * x(t)$ for the following values of T : (a) $T = 3$, (b) $T = 2$, (c) $T = 1.5$.

Fig. 2-10

$$y(t) = h(t) * \delta_T(t) = h(t) * \left[\sum_{n=-\infty}^{\infty} \delta(t - nT) \right]$$

$$= \sum_{n=-\infty}^{\infty} h(t) * \delta(t - nT) = \sum_{n=-\infty}^{\infty} h(t - nT) \quad (2.69)$$

(a) For $T = 3$, Eq. (2.69) becomes

$$y(t) = \sum_{n=-\infty}^{\infty} h(t - 3n) \quad T > w$$

which is sketched in Fig. 2-11(a).

(b) For $T = 2$, Eq. (2.69) becomes

$$y(t) = \sum_{n=-\infty}^{\infty} h(t - 2n) \quad T = w$$

which is sketched in Fig. 2-11(b).

(c) For $T = 1.5$, Eq. (2.69) becomes

$$y(t) = \sum_{n=-\infty}^{\infty} h(t - 1.5n) \quad T < w$$

which is sketched in Fig. 2-11(c). Note that when $T < 2$, the triangular pulses are no longer separated and they overlap.

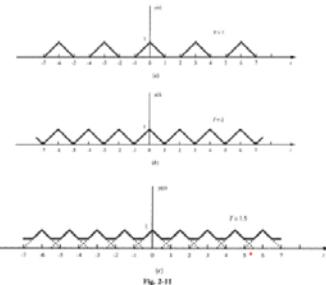
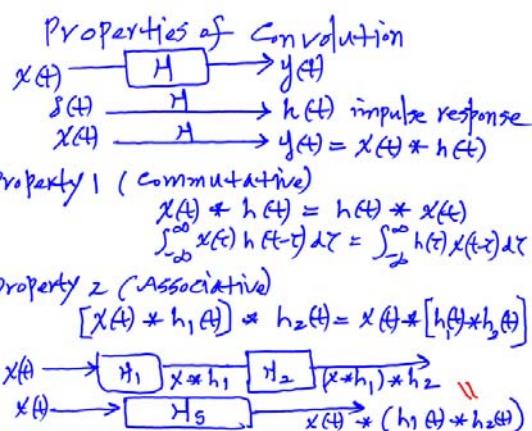
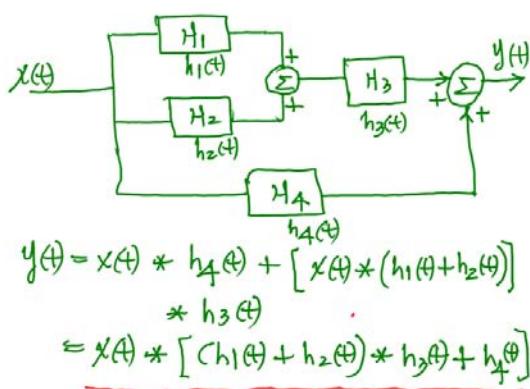


Fig. 2-11



Property 3 (Distributive)

$$\begin{aligned} x(t) * h_1(t) + x(t) * h_2(t) &= x(t) * [h_1(t) + h_2(t)] \\ x(t) \xrightarrow{H_1} x(t) * h_1(t) &+ x(t) \xrightarrow{H_2} x(t) * h_2(t) \\ &\xrightarrow{+} x(t) * [h_1(t) + h_2(t)] \end{aligned}$$



General Form of n^{th} -order Linear (Ordinary) differential equation

$$a_n \frac{d^n y(t)}{dt^n} + a_{n-1} \frac{d^{n-1} y(t)}{dt^{n-1}} + \dots + a_1 \frac{dy(t)}{dt} + a_0 y(t)$$

$$= b_m \frac{d^m x(t)}{dt^m} + b_{m-1} \frac{d^{m-1} x(t)}{dt^{m-1}} + \dots + b_1 \frac{dx(t)}{dt} + b_0 x(t)$$

where $a_k, k=0, n$ and $b_j, j=0, m$ are constants

More compact form

$$\sum_{k=0}^n a_k \frac{d^k y(t)}{dt^k} = \sum_{j=0}^m b_j \frac{d^j x(t)}{dt^j} \quad (\text{I})$$

solution $y(t) = y_c(t) + y_p(t)$

$y_c(t)$ = complementary solution (natural response)
 $y_p(t)$ = particular solution (forced response)

Natural response $= y(t)$ when $x(t) = 0$, i.e. zero input

From (A) $\sum_{k=0}^n \alpha_k \frac{d^k y}{dt^k} = 0$ (II)

In general $y_c(t) = C_1 e^{\zeta_1 t}$ (III)

From (II) & (III), we obtain

$(\sum_{k=0}^n \alpha_k \zeta_1^k) C_1 e^{\zeta_1 t} = 0 \Rightarrow$

$$\Rightarrow \alpha_n \zeta_1^n + \alpha_{n-1} \zeta_1^{n-1} + \dots + \alpha_1 \zeta_1 + \alpha_0 = 0$$

$$\alpha_n (\zeta - \zeta_1)(\zeta - \zeta_2) \dots (\zeta - \zeta_n) = 0 \quad (\text{III})$$

then $y_c(t) = C_1 e^{\zeta_1 t} + C_2 e^{\zeta_2 t} + \dots + C_n e^{\zeta_n t}$

(Example)

$$+ \frac{R}{C} \frac{dy}{dt} + y = x \quad \text{RC} \frac{dy}{dt} + y = x$$

$$\alpha_1 \frac{dy}{dt} + \alpha_0 y = x$$

$$\alpha_1 = RC, \alpha_0 = 1, b_0 = 1$$

$$y_c(t) = ? \quad \text{set } x(t) = 0$$

$$RC \frac{dy}{dt} + y = 0$$

Let $y_c(t) = C_1 e^{\zeta_1 t}$

$$RC \frac{1}{\zeta_1} (C_1 e^{\zeta_1 t}) + C_1 e^{\zeta_1 t} = 0$$

$$RC C_1 \zeta_1 e^{\zeta_1 t} + C_1 e^{\zeta_1 t}$$

$$= (RC \zeta_1 + 1) C_1 e^{\zeta_1 t} = 0$$

$$\Rightarrow RC \zeta_1 + 1 = 0 \Rightarrow \zeta_1 = -\frac{1}{RC}$$

$$y_c(t) = C_1 e^{-\frac{1}{RC}t}$$

$$x(t) \quad \begin{cases} + & t > 0 \\ 0 & t < 0 \end{cases} \quad y(0) = 0 \quad \text{RC} \frac{dy}{dt} + y(t) = x(t)$$

For $x(t) = u(t) = \begin{cases} 1 & t > 0 \\ 0 & t < 0 \end{cases}$

$$RC \frac{dy}{dt} + y_p(t) = 1, t > 0$$

$$y_p(t) = P(\text{constant}), \frac{dy_p}{dt} = 0$$

$$y_p(t) = P = 1$$

$$y(t) = y_c(t) + y_p(t) = C_1 e^{-\frac{1}{RC}t} + 1$$

$$y(0) = 0 = C_1 e^{-\frac{1}{RC}(0)} + 1 \Rightarrow C_1 = -1$$

$$\Rightarrow y(t) = 1 - e^{-\frac{1}{RC}t}$$

Review of BIBO stability

$$y(t) = H x(t)$$

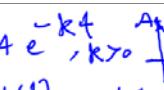
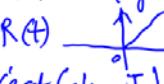
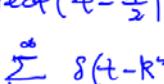
$$= \int_{-\infty}^{\infty} x(\tau) h(t-\tau) d\tau$$

Taking absolute value of both sides

$$|y(t)| = \left| \int_{-\infty}^{\infty} x(\tau) h(t-\tau) d\tau \right| \leq \int_{-\infty}^{\infty} |x(\tau)| |h(t-\tau)| d\tau$$

$|y(t)| \leq N \text{ for all } |x(t)| \leq M \quad (\text{B.I.B.O.})$

If and only if $\int_{-\infty}^{\infty} |h(\tau)| d\tau < \infty$

- $h(t) = A e^{-kt}$  BIBO
- $h(t) = u(t)$ Integrator, not BIBO
- $h(t) = R(t)$  not BIBO
- $h(t) = \text{rect}(t - \frac{T}{2}) T$  BIBO
- $h(t) = \sum_{k=0}^{\infty} \delta(t - kT)$  not BIBO

